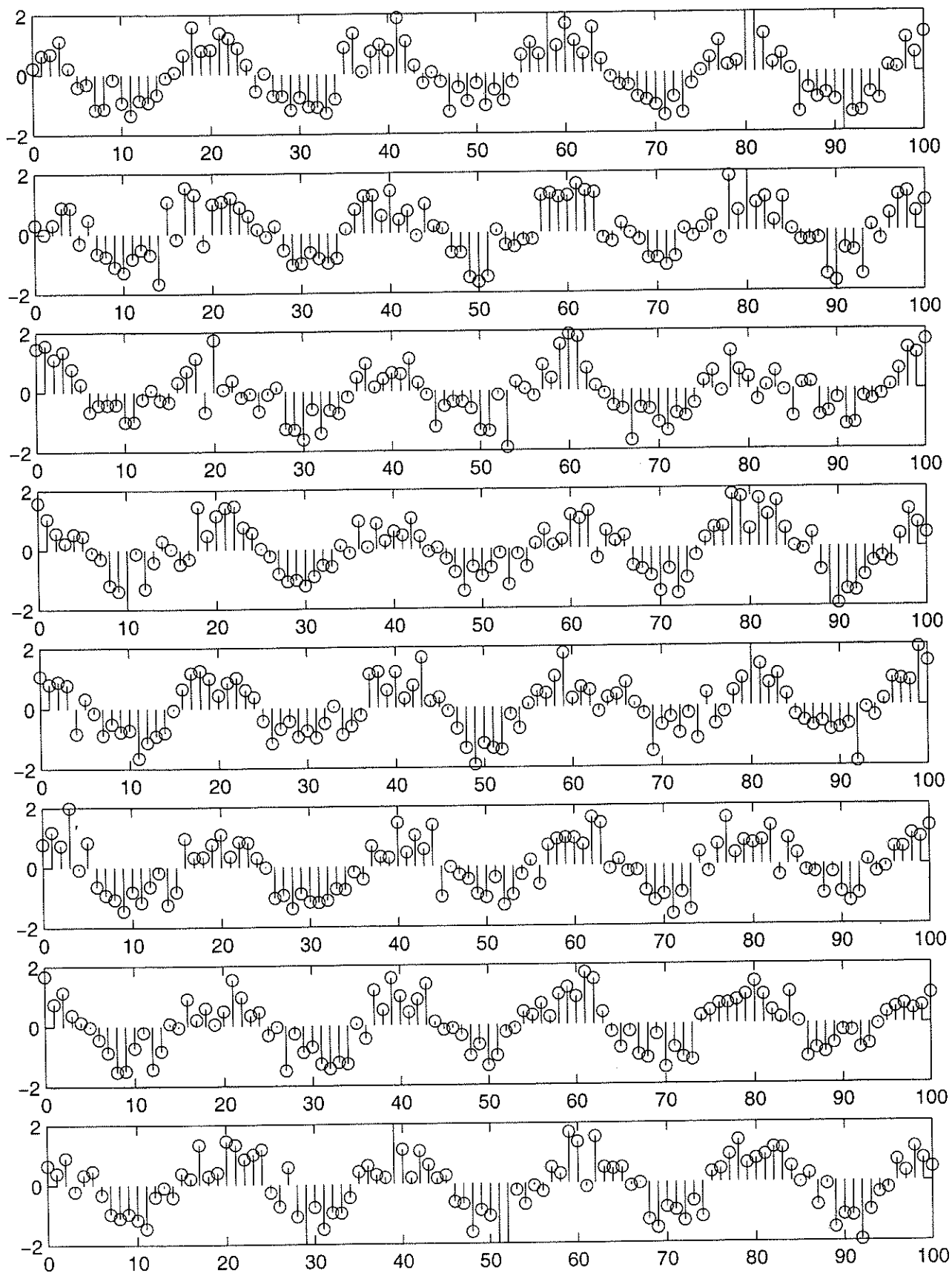


Discrete-Time Random Signals

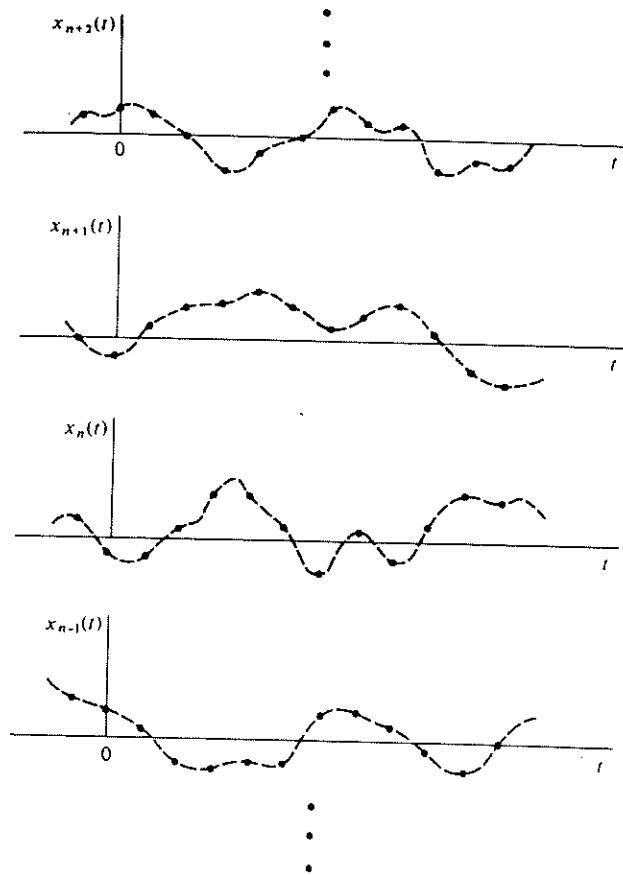
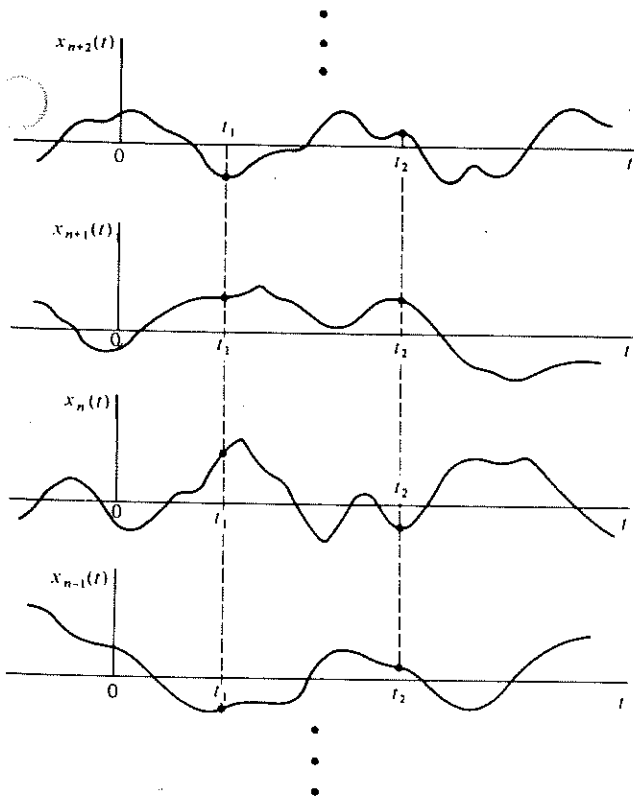
Classic Example: Sinusoid + Noise



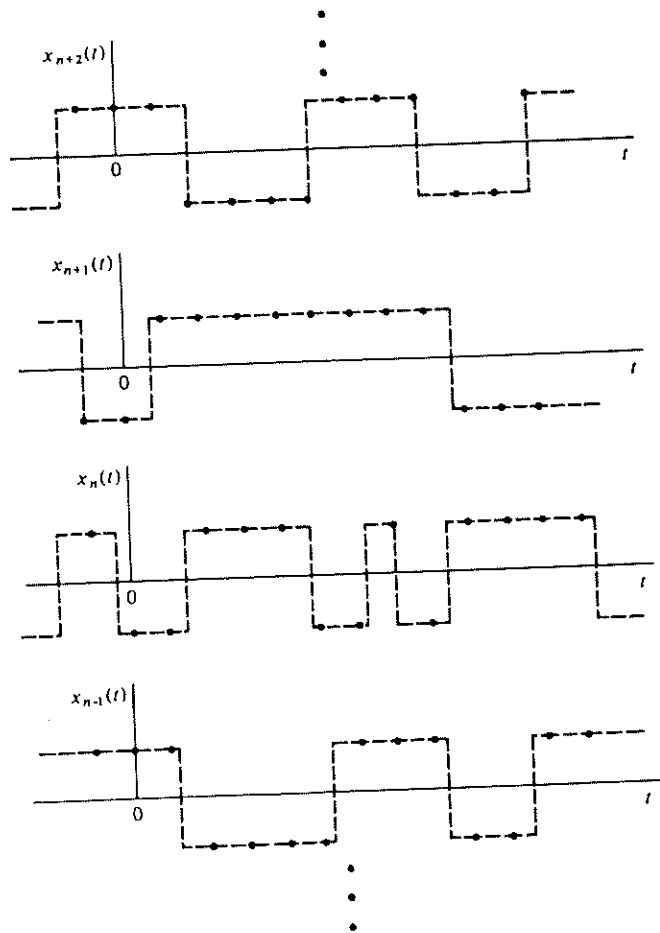
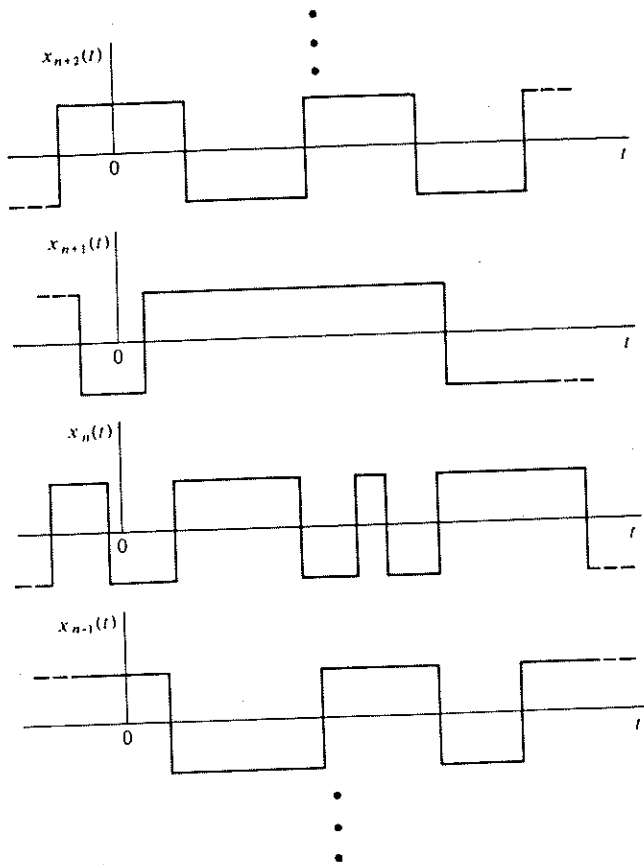
Definition: A random process

is a family or ensemble of signals corresponding to every possible outcome of a certain signal measurement or experiment. Each signal in the ensemble is called a "realization" of the process.

Ex.



Ex. Random Binary Process

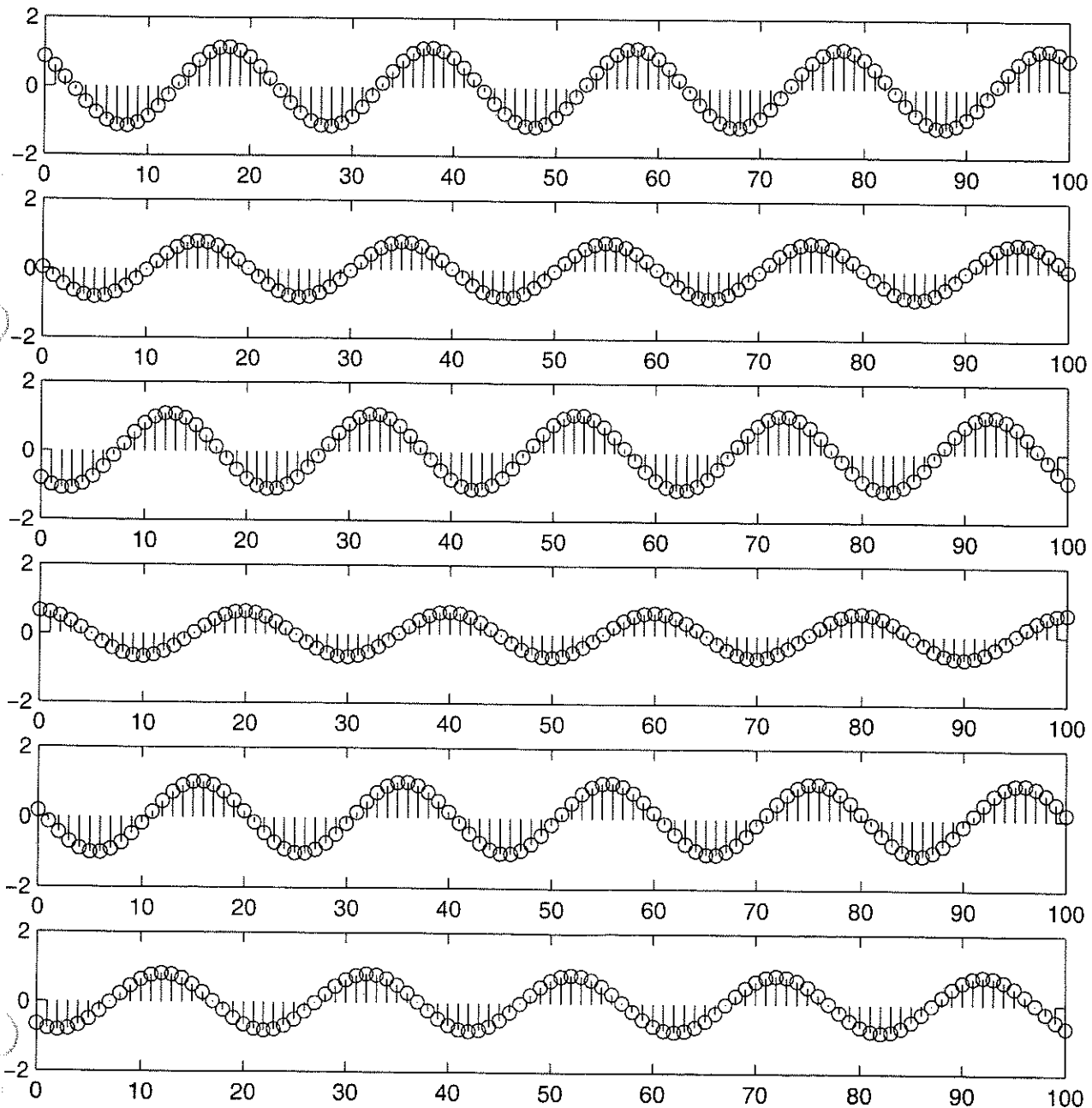


Ex. Random Sinusoidal Process

$$X[n] = A \cos(\omega_0 n + \phi)$$

A, ω_0, ϕ may all be
random variables

random
amplitude
& phase



Definition: The mean of a random process is the average of all realizations of the process.

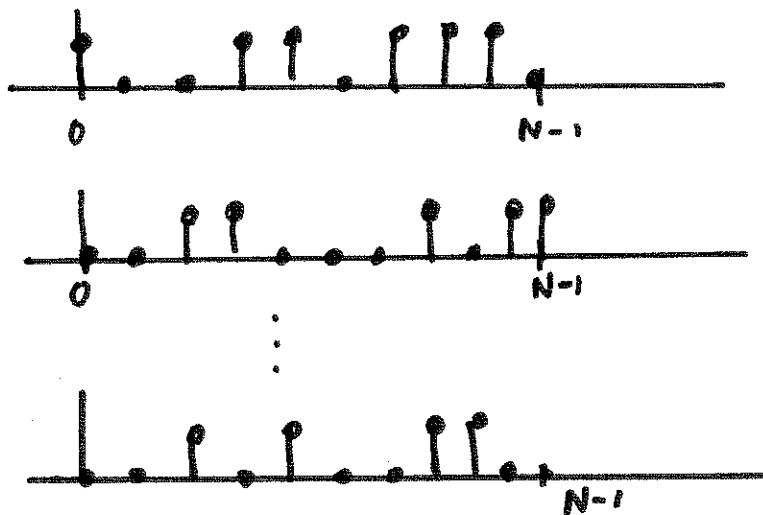
Ex. N-pt binary random sequence

$$X[n] = \begin{cases} 1, & \text{with prob } p \\ 0, & \text{with prob } (1-p) \end{cases} \quad n=0, \dots, N-1$$

independent

$$\Pr(X[n]=1) = p \quad \Pr(X[n]=0) = 1-p$$

Realizations:



What is $\text{mean}(X[n])$?

The mean is also known as the expectation, and is denoted by

$$m_x[n] = E[x[n]]$$

↑
expectation operator

⇒ take average of all possible realizations of $x[n]$

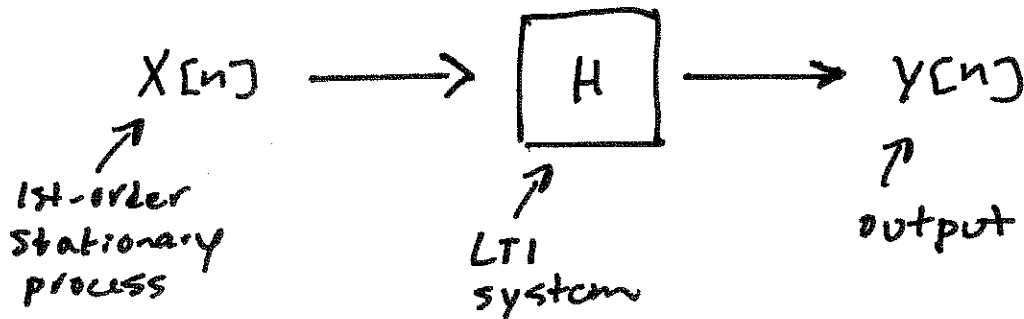
First-order Stationary Processes

A process is 1st-order stationary if

$$m_x[n] = m_x \text{ (a constant independent of } n \text{)}$$

Ex. N-pt binary process revisited

First-order Stationarity and LTI Systems



LTI system is deterministic, but input is random.

What about output?

$$m_Y[n] = E \left[\sum_{k=-\infty}^{\infty} h[k] x[n-k] \right]$$

=

Definition: The autocorrelation function

of a random process is the average product of a signal realization with a time-shifted version of itself.

$$R_{xx}[n, n+m] = E[x[n]x[n+m]]$$

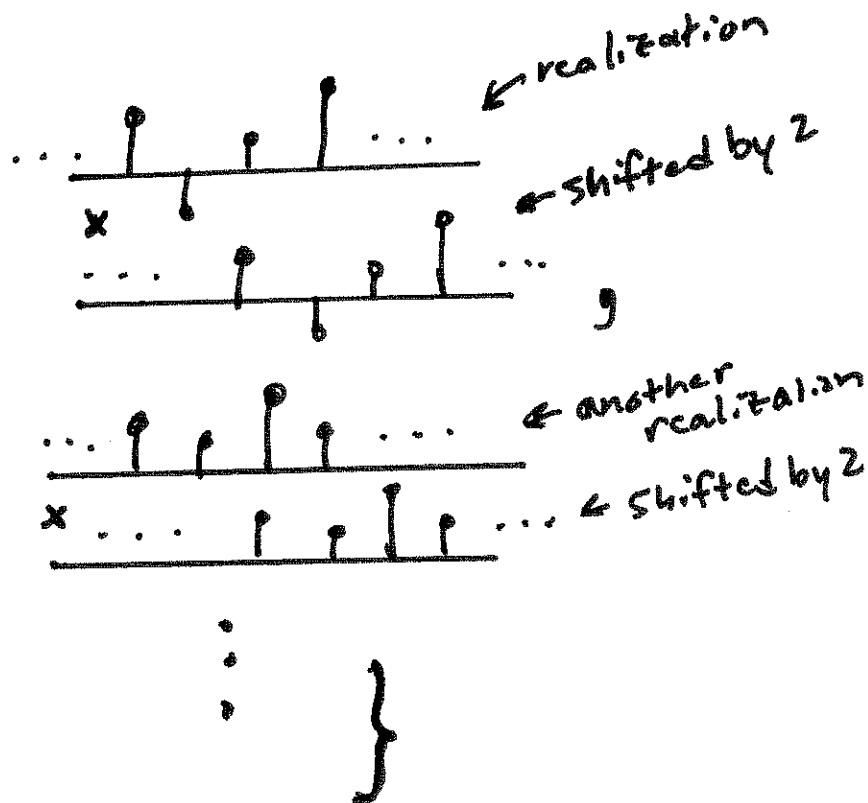
↑ autocorrelation function

↑ Average over all possible realizations

Ex.

$$R_{xx}[n, n+2] =$$

Average {



Ex. Random Binary Process

$$\text{Assume } \Pr(x[n] = 1) = \Pr(x[n] = -1) = \frac{1}{2}$$

$$\Rightarrow m_x[n] =$$

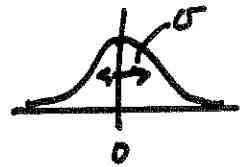
$$R_{xx}[n, n+m] = E[x[n] x[n+m]]$$

Ex. Gaussian White Noise (GWN)

"white" \Rightarrow $x[n]$ are independent
and zero-mean

$$\Pr [a \leq x[n] \leq b] = \int_a^b \underbrace{\frac{1}{\sqrt{2\pi\sigma^2}} e^{-x^2/2\sigma^2}}_{\text{Gaussian density function}} dx$$

Gaussian density
function
 $\sigma^2 = \text{variance}$



$$m_x = E[x[n]] = \int_{-\infty}^{\infty} x \cdot \frac{1}{\sqrt{2\pi\sigma^2}} e^{-x^2/2\sigma^2} dx = 0$$

$$R_{xx}[n, n+m] = E[x[n] x[n+m]]$$

=

Second-order Stationary Processes

A random process is 2nd-order stationary if

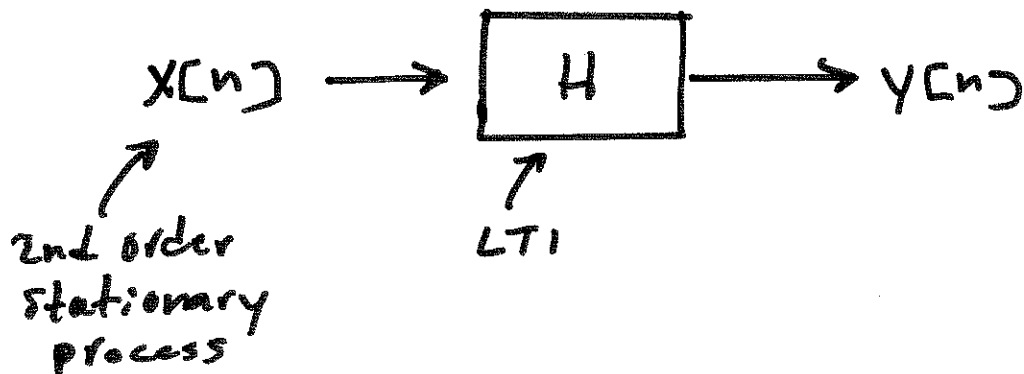
$$R_{xx}[n, n+m] \equiv R_{xx}[m]$$

That is, the autocorrelation function only depends on m , the shift, and not on n .

Ex. Binary Process

Ex. GWN

Stationary Inputs & LTI Systems



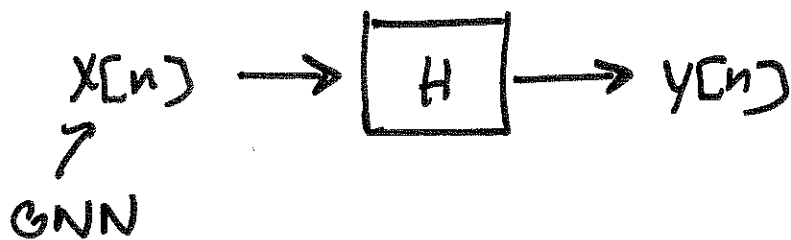
$$\begin{aligned} m_Y[n] &= E \left[\sum_{k=-\infty}^{\infty} h[k] X[n-k] \right] \\ &= \sum_{k=-\infty}^{\infty} h[k] E[X[n-k]] = m_X \sum_{k=-\infty}^{\infty} h[k] \end{aligned}$$

$$\begin{aligned} R_{YY}[n, n+m] &= E \left[\sum_{k=-\infty}^{\infty} h[k] X[n-k] \cdot \sum_{r=-\infty}^{\infty} h[r] X[n+m-r] \right] \\ &= \end{aligned}$$

$$\Rightarrow R_{YY}[m] = \sum_{l=-\infty}^{\infty} R_{XX}[m-l] \cdot c_{hh}[l]$$

where $c_{hh}[l] = \sum_{k=-\infty}^{\infty} h[k] h[l+k]$

Ex. GWN into an LTI System



$$m_y = m_x \sum_{k=-\infty}^{\infty} h[k] = 0$$

$$R_{yy}[m] = \sum_{\ell=-\infty}^{\infty} R_{xx}[m-\ell] \cdot \sum_{k=-\infty}^{\infty} h[k] h[\ell+k]$$

$\nearrow \sigma^2 \delta[m-\ell]$

$$= \sum_{\ell=-\infty}^{\infty} \sigma^2 \delta[m-\ell] \cdot \sum_{k=-\infty}^{\infty} h[k] h[\ell+k]$$

$$= \sum_{k=-\infty}^{\infty} \sigma^2 h[k] h[m+k]$$

Frequency Domain Analysis

(assume 2nd-order stationarity)

Power Spectrum

$$S_{xx}(\omega) = \sum_{m=-\infty}^{\infty} R_{xx}[m] e^{-j\omega m} \quad (1)$$

$$R_{xx}[m] = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_{xx}(\omega) e^{j\omega m} d\omega \quad (2)$$

① & ② are called the Wiener-Khinchin Relations

Frequency Domain Analysis of LTI Systems

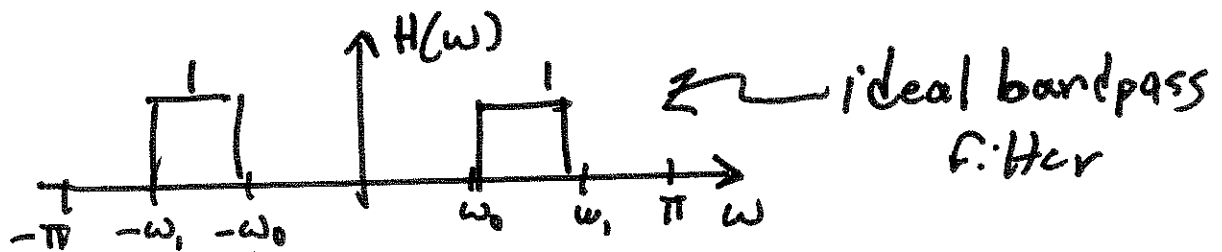
$$R_{yy}[m] = \sum_{l=-\infty}^{\infty} R_{xx}[m-l] C_{hh}[l]$$

$$\Rightarrow S_{yy}(\omega) = |H(\omega)|^2 \cdot S_{xx}(\omega)$$

↑
Power spectrum of $x[n]$ is
"shaped" by $|H(\omega)|^2$

Ex.

$$H(\omega) = \begin{cases} 1, & \omega_0 \leq |\omega| \leq \omega_1 \\ 0, & \text{otherwise} \end{cases}$$



$$S_{yy}(\omega) = |H(\omega)|^2 S_{xx}(\omega)$$

In particular, the average output power is

$$E[\bar{z}^2[n]] = R_{yy}[0] = \frac{1}{2\pi} \int_{-\pi}^{\pi} S_{yy}(\omega) d\omega$$

$$= \frac{1}{2\pi} \int_{-\omega_1}^{-\omega_0} S_{xx}(\omega) d\omega + \frac{1}{2\pi} \int_{\omega_0}^{\omega_1} S_{xx}(\omega) d\omega$$

This shows that we can interpret the area under $S_{xx}(\omega)$ for $\omega_0 \leq |\omega| \leq \omega_1$

as the average input power in

that frequency band. Thus, $S_{xx}(\omega)$ can be viewed as a density function for power in the spectral (freq) domain.

Applications

Ex. Noise Removal

Suppose that we make noisy measurements of a random signal $s[n]$ in GWN $w[n]$:

$$x[n] = s[n] + w[n]$$

signal:

$$m_s = 0$$

$$R_{ss}[m], S_{ss}(\omega)$$

arbitrary

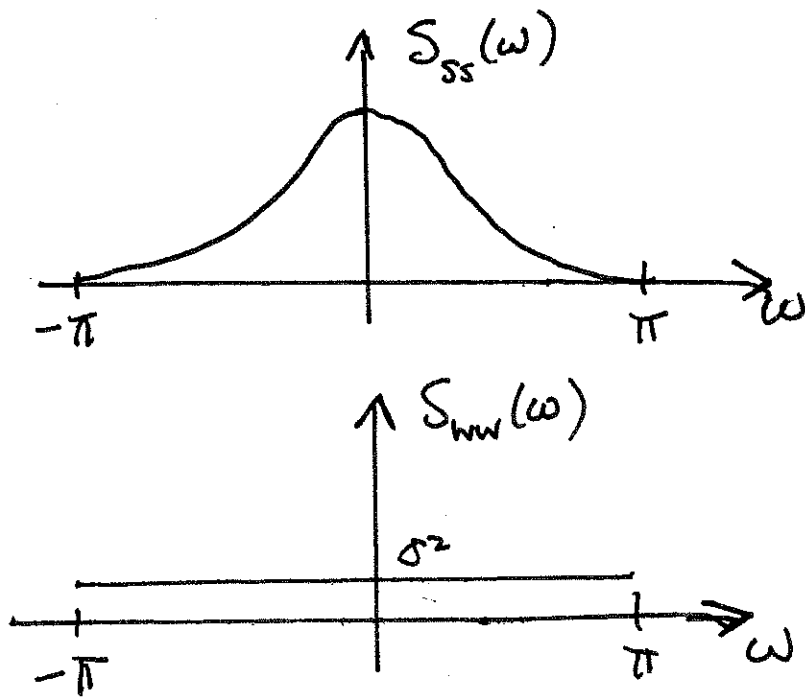
noise:

$$m_w = 0$$

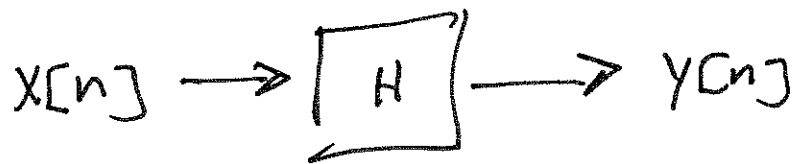
$$R_{ww}[m] = \sigma^2 \delta[m]$$

$$S_{ww}(\omega) = \sigma^2 \text{ const.}$$

Consider the case



Define the noise removal filter

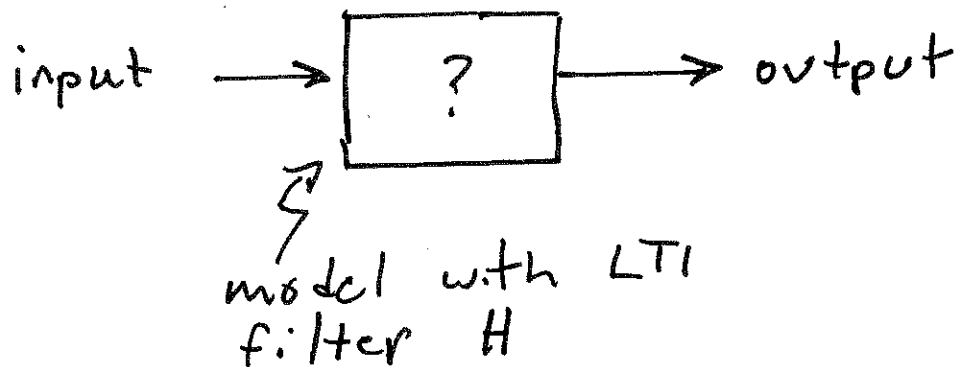


$$H(\omega) = \begin{cases} 1, & |\omega| < \omega_c \\ 0, & \omega_c \leq |\omega| \leq \pi \end{cases}$$

How to choose ω_c ?

Ex. System Identification

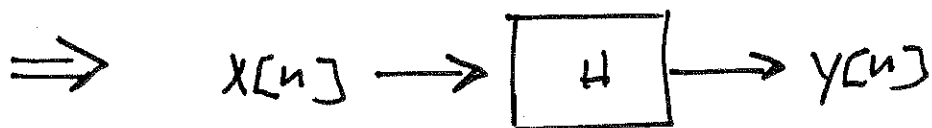
Suppose that we wish to model an unknown physical system with an LTI filter.



How can we "identify" the filter H ?

White Noise Probing:

input $\equiv x[n] = \text{GWN}$



$$m_y = 0$$

$$R_{yy}[m] = \sum_{k=-\infty}^{\infty} \sigma^2 h[k] h[m+k]$$

$$S_{yy}(\omega) = \sigma^2 |H(\omega)|^2$$

← close, but only tells us magnitude

Stationary Processes

○ Properties of Correlation Function:

$$(1) \quad |R_{xx}[m]| \leq R_{xx}[0]$$

$$(2) \quad R_{xx}[-m] = R_{xx}[m]$$

(3) power of process $x[n]$ is given by $R_{xx}[0]$

proof of (1): $(x[n] - x[n+m])^2 \geq 0$
 $\Rightarrow E[(x[n] - x[n+m])^2] \geq 0$

similarly, $(x[n] + x[n+m])^2 \geq 0$
 $\Rightarrow E[(x[n] + x[n+m])^2] \geq 0$

proof of (2):

$$R_{xx}[m] = E[x[n]x[n+m]]$$
$$R_{xx}[-m] = E[x[n]x[n-m]]$$

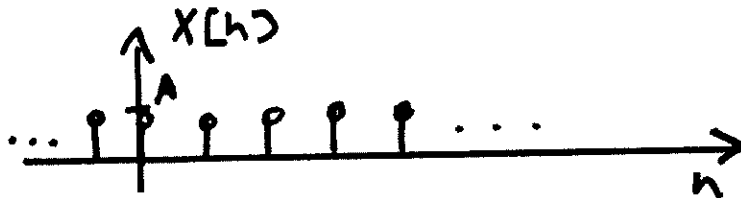
take $n = n' - m$

$$R_{xx}[m] = E[x[n'-m]x[n']] = R_{xx}[-m]$$

for any $n!$

Correlation

Total Correlation :



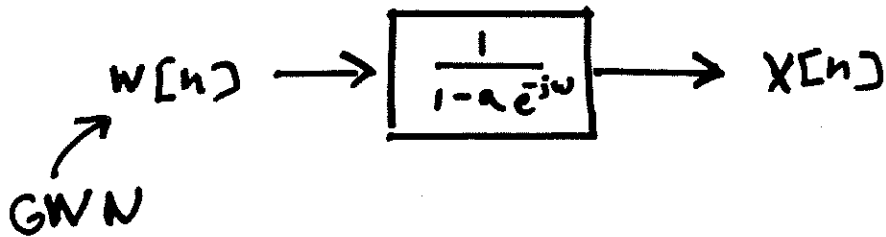
$$X[n] = A, \quad P_A(a) = \frac{1}{\sqrt{2\pi\sigma^2}} e^{-a^2/2\sigma^2}$$

mean :

correlation function :

Zero Correlation :

Strong Correlation:



$$X[n] = a X[n-1] + w[n] ; 0 < a < 1$$

impulse response $h[n] =$

mean:

Correlation function:

Stationarity

Let $x[n]$ be a random process.

Then we can talk about

$$\Pr(|x[n]| < a)$$

or $\Pr(a < x[n] < b)$.

We can also study the joint probability distributions for two or more samples:

$$\Pr(a < x[n] < b, c < x[n+m] < d)$$

or $\Pr(|x[n]| \leq a, |x[n+1]| \leq b, |x[n+2]| \leq c)$

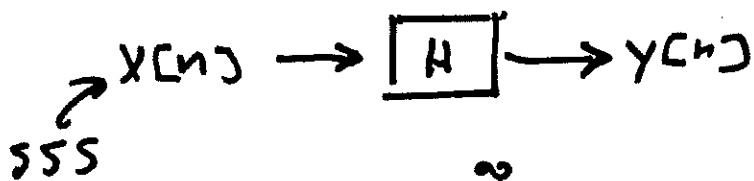
Strict-Sense Stationarity (SSS)

Strict-sense stationarity means that these joint probability distributions are time-invariant.

ex. SSS

$$\begin{aligned} \Pr(a < x[n] < b, c < x[n+m] < d) \\ = \Pr(a < x[n+k] < b, c < x[n+k+m] < d) \\ \text{for every } k \in \mathbb{Z}. \end{aligned}$$

SSS \neq LTI systems



$$y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

$$\begin{aligned} \Pr(a \leq y[n] \leq b) &= \Pr\left(a \leq \sum_k h[k] x[n-k] \leq b\right) \\ &= \Pr\left(a \leq \sum_k h[k] x[n+m-k] \leq b\right) \quad \text{by SSS of } x[n] \\ &= \Pr(a \leq y[n+m] \leq b) \end{aligned}$$

Wide-Sense Stationary (WSS)

$$(1) E[x[n]] = m_x \text{ const.}$$

$$(2) E[x[n]x[n+m]] = R_{xx}(m)$$

(1) \Rightarrow average value of process
is time-invariant

(2) \Rightarrow average power of process
and correlation between samples
is time-invariant

WSS is weaker than SSS:

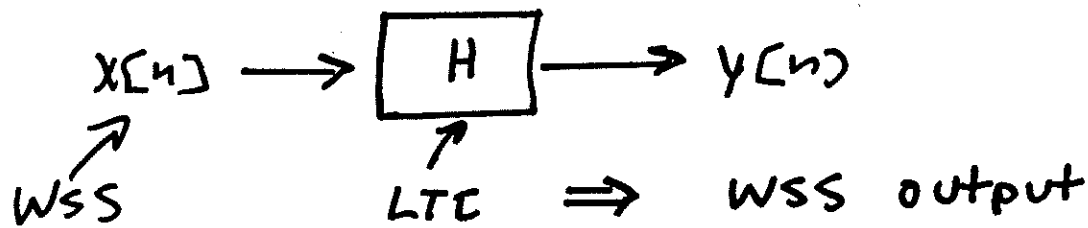
$$\text{SSS} \Rightarrow \text{WSS}$$

~~\Leftarrow~~

WSS tells us that average behavior
of single samples and products of
samples is time-invariant.

SSS tells us that the joint probability
distributions for arbitrary collections
of samples are time-invariant.

WSS & LTI Systems



In most cases and applications WSS is sufficient for analysis purposes.

$$(1) E[x[n]] = m_x \Rightarrow E[y[n]] = m_y \text{ const.}$$

$$(2) E[x[n]x[n+m]] = R_{xx}[m]$$

Correlation structure is sufficient to describe frequency content of input

$$S_{xx}(\omega) = \sum_{m=-\infty}^{\infty} R_{xx}[m] e^{-j\omega m}$$

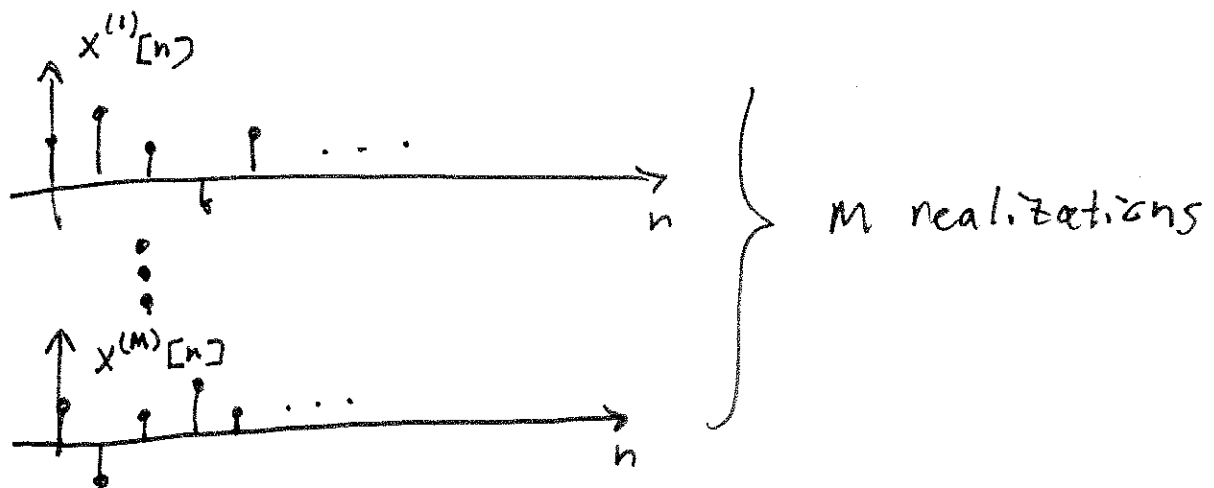
Thus, WSS is sufficient for frequency-domain analysis of random processes through LTI systems.

Estimating Means and Autocorrelations from Data

Suppose that we make several independent observations of a random process:

$$x^{(i)}[n], \quad i = 1, \dots, M$$

realizations



We can estimate the mean function using

$$\hat{m}_x[n] = \frac{1}{M} \sum_{i=1}^M x^{(i)}[n]$$

$$\hat{m}_x[n] \rightarrow m_x[n] \quad \text{as } M \rightarrow \infty$$

We can estimate the autocorrelation function according to

$$\hat{R}_{xx}[n, n+m] = \frac{1}{M} \sum_{i=1}^M x^{(i)}[n] x^{(i)}[n+m]$$

If the process is stationary, then we can estimate the mean and autocorrelation from a single realization using time-averaging.

$$\hat{m}_x = \frac{1}{N} \sum_{n=0}^{N-1} x[n]$$
$$\hat{R}_{xx}[m] = \frac{1}{N-m} \sum_{n=0}^{N-m-1} x[n] x[n+m]$$

$$\hat{m}_x \rightarrow m_x \quad \text{as } N \rightarrow \infty$$
$$\hat{R}_{xx}[m] \rightarrow R_{xx}[m]$$